Semiconductor photon detectors

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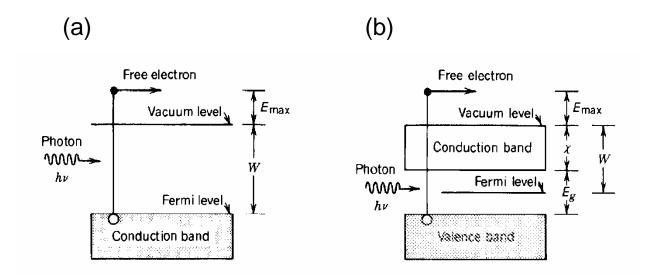
Fundamentals of Photonics B.A. Saleh, M.C. Teich 1991, John Wiley & Sons, Inc.

Classes of photodetectors

Thermal detectors: photon energy converted into heat

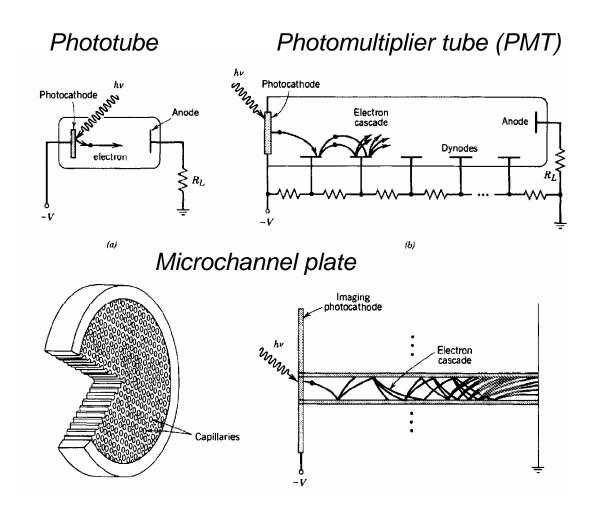
Photoelectric detectors: photon energy converted into mobile charge carries to yield electric current

External photoeffect: photoelectric emission

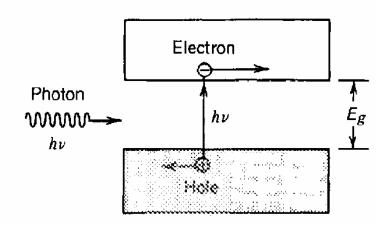


Photoelectric emission from (a) a metal and (b) a semiconductor

Photodetectors based on photoelectric emission



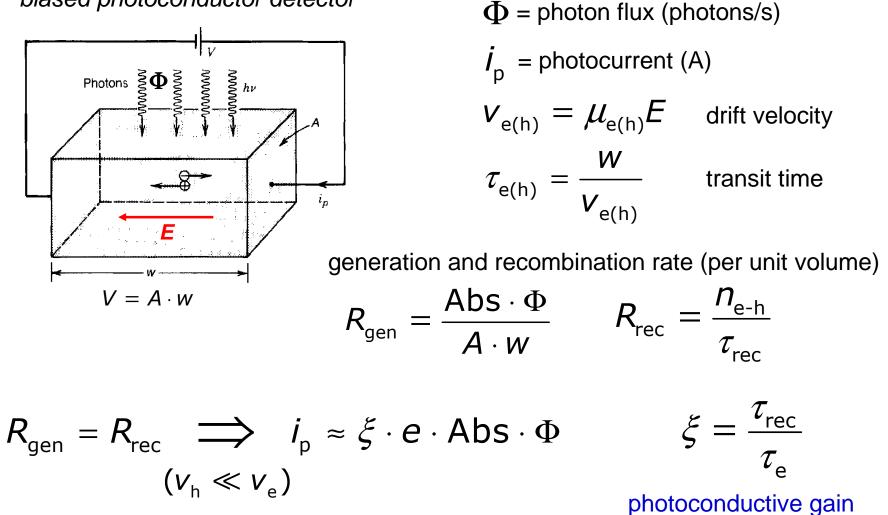
Internal photoeffect: photoconductivity



electron-hole photogeneration in a semiconductor

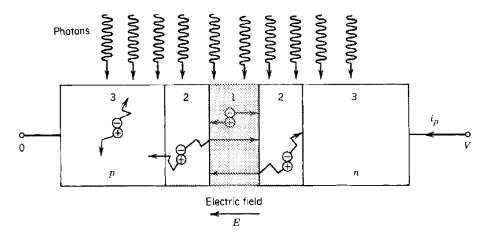
Photoconductor detectors

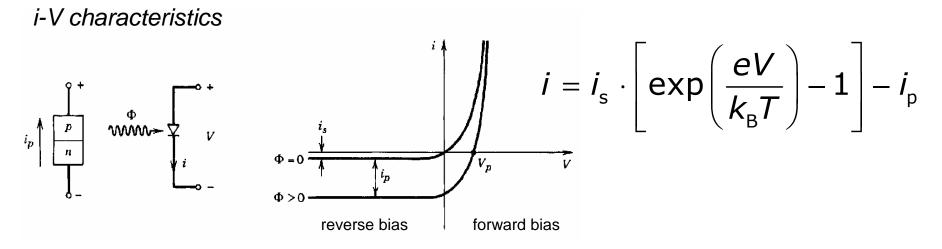
biased photoconductor detector



Photodiodes

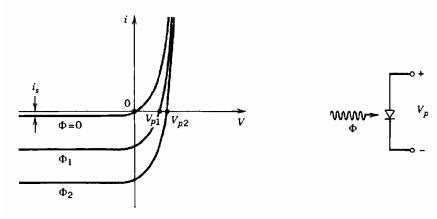
Electron-hole pair generation in a p-n photodiode



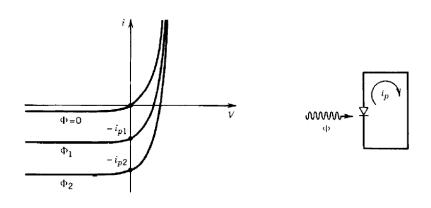


Operation of a *p*-*n* photodiode

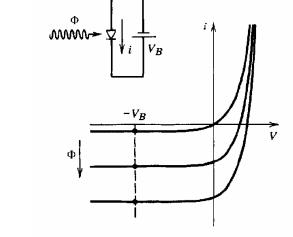
(i) open-circuit (photovoltaic) operation



(ii) short-circuit operation

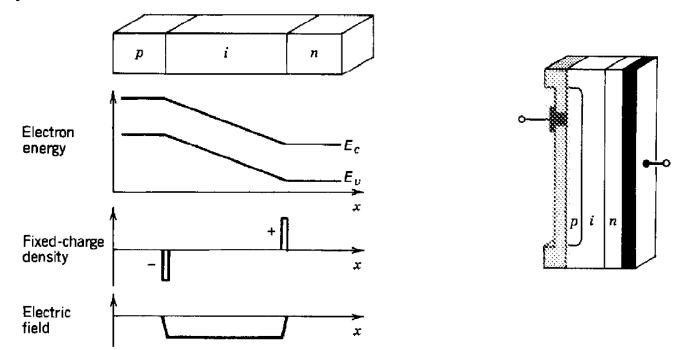


(iii) reverse-bias (**photoconductive**) operation



The *p-i-n* photodiode

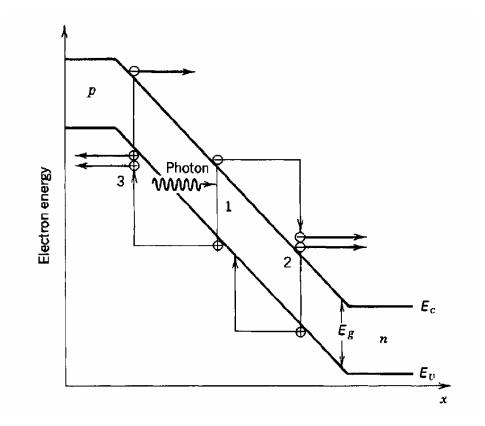
Intrinsic (lightly doped) layer sandwitched between the *n* and *p* layers



Heterostructure devices: AlGaAs/GaAs, InGaAs/InP, HgCdTe/CdTe ...

The avalanche photodiode

An avalanche photodiode (APD) converts each detected photon into a cascade of moving carrier pairs by *impact ionization*



Properties of semiconductor photodetectors

- Quantum efficiency
- Responsivity
- Gain
- Response time

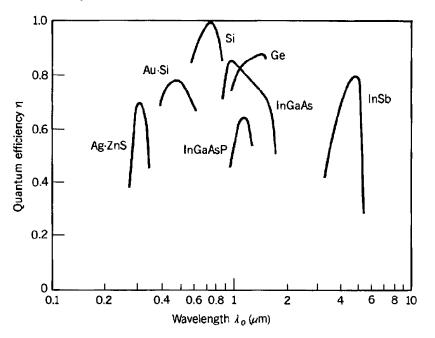
Quantum efficiency

Probability that a single photon incident on the device generates a photocarrier pair that contribute to the detector current

$$\eta = \xi \left[1 - \exp(-\alpha d) \right] (1 - R) = \xi \cdot Abs$$

- R = optical reflectance at the surface
- α = absorption coefficient of the material
- d = photodetector depth
- ξ = e-h fraction contributing to the detector current

quantum efficiency vs. wavelength for photodiodes based on various materials



Responsivity

Relates the electric current (i_p) flowing in the device to the incident optical power (P)

$$i_{p} = \eta \frac{eP}{hv} = \Re P \implies \Re = \eta \frac{e}{hv} = \eta \frac{e\lambda_{0}}{hc} \left(\frac{A}{W}\right)$$

$$P = \text{incident optical power}$$

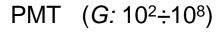
$$hv = \text{photon energy}$$

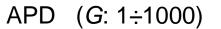
$$I_{0} = \frac{10}{0.8} = \frac{10}{0.4} = \frac{10}{0.5} = \frac{10}{10} = \frac{10}{$$

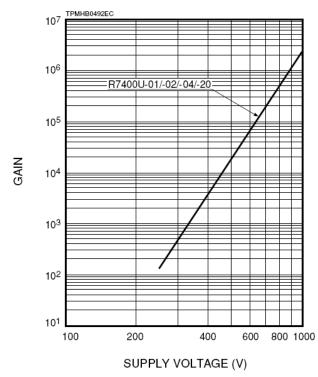
Device with gain

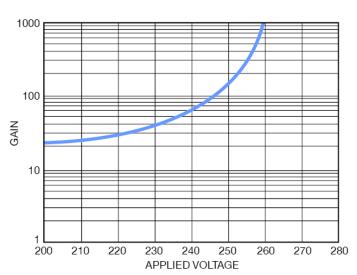
Each carrier pair produces in the external circuit a current pulse of charge q > e:

$$G = \frac{q}{e} \implies \Re = G\eta \frac{e\lambda_0}{hc}$$



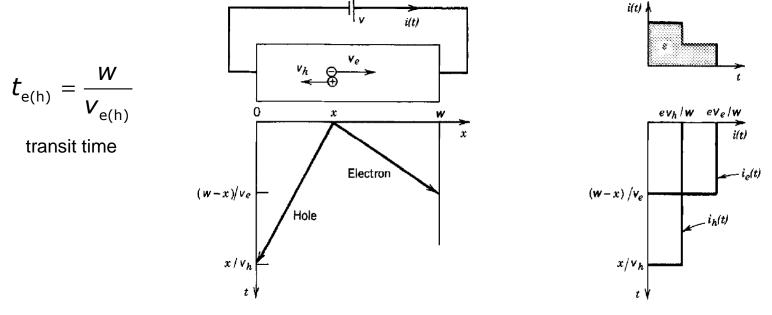






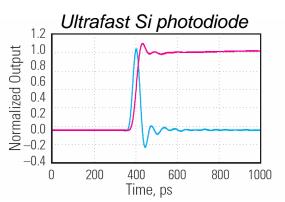
Response time

Charge delivered to the external circuit by carrier motion occupies an extended time



Impulse-response function determined by convolving i(t) with time-constant spread function f(t):

$$f(t) = \frac{1}{RC} \exp(-\frac{t}{RC})$$



Noise in photodetectors

Photocurrent randomly fluctuates above and below its average:

$$i_{\rm p} = \Re P$$
 $i_{\rm p} = \overline{i_{\rm p}} + \delta i_{\rm p}$ $\sigma_{\rm p}^2 = \langle \delta i_{\rm p}^2 \rangle = \sum_k \sigma_k^2$

Several noise sources:

- Photocurrent noise. Reflects photon-counting statistics
- Gain noise. Induced by randomness of the gain process
- Receiver circuit noise. Various components of the photodetector, such as resistors and transistors, give rise to fluctuations of the electric current

Photocurrent noise

Number of photoelectrons collected in time interval T

$$\overline{m} = \eta \Phi T = \frac{\eta \Phi}{2B}$$
 $T = \frac{1}{2B}$ (*B* = detection bandwidth)

Photoelectron noise (reflecting Poisson distribution for random counting)

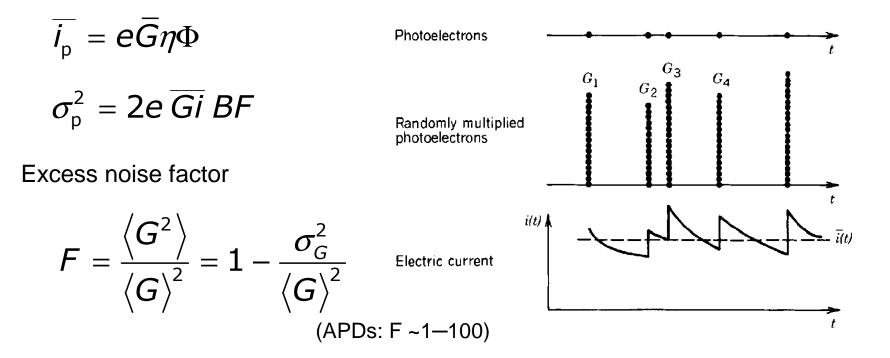
$$\sigma_m^2 = \overline{m}$$
Photons
Photoelectrons
Current pulses
$$\overline{i_p} = \frac{e}{T} \overline{m} = \eta e \Phi$$
Photons
$$\lim_{t \to 1^+} \frac{1}{i_{p}} = \frac{e}{T} \overline{m} = \eta e \Phi$$
Photocurrent noise
$$\sigma_p^2 = \frac{e^2}{T^2} \sigma_m^2 = 2e\overline{iB}$$
Dark current noise
$$\sigma_i^2 = 2e \overline{i_d} B$$

Gain noise

Randomness of gain

$$G = \overline{G} + \delta G$$
 $\sigma_G^2 = \left< \delta G^2 \right>$

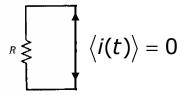
Photocurrent mean and variance



Circuit noise: thermal noise

Thermal noise (or *Johnson noise*) arises from random motion of mobile carriers in resistive electrical materials at finite temperatures

Random electric current *i*(*t*) in the absence of an external power source



Power spectral density of thermal noise

$$S_{\rm th}(f) = \frac{4}{R} \frac{hf}{\exp\left(\frac{hf}{k_{\rm B}T}\right) - 1}$$

Noise variance

$$\sigma_i^2 = \int_0^B S_{\rm th}(f) df \approx \frac{4k_{\rm B}T}{R}B$$

$$S_{\rm th}(f) \approx rac{4k_{\rm B}T}{R}$$

for $f \ll k_{\rm B}T / h = 6.25 \,\text{THz}$ (300K)

Signal-to-noise ratio

Signal-to-noise ratio (SNR)

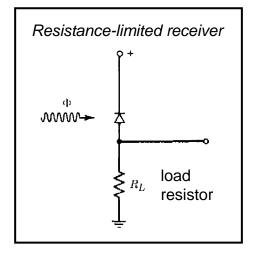
$$SNR = \frac{i_{p}^{2}}{\sigma_{i}^{2}} = \frac{i_{p}^{2}}{\sigma_{p}^{2} + \sigma_{r}^{2}} \qquad i_{p} = e\overline{G}\eta\Phi \qquad photocurrent mean$$

$$\sigma_{p}^{2} = 2e\overline{G}i_{p}FB \qquad photocurrent variance$$

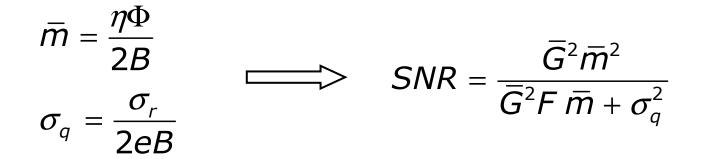
Receiver noise

$$\sigma_r^2 = \left[2e\,\overline{Gi_d}\,F + \frac{4k_{\rm B}T}{R_{\rm L}}\right]B + \int_0^B S_{\rm add}(f)df$$

dark current noise + thermal noise + additional noise sources

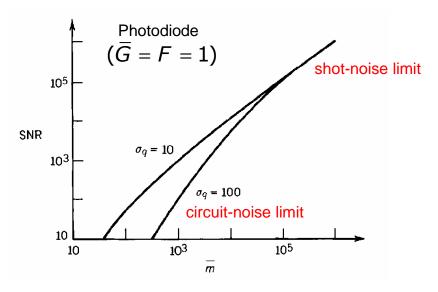


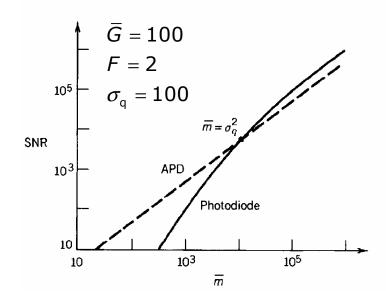
Signal-to-noise ratio



SNR degradation with increasing receiver noise

Effect of gain on SNR





Noise-equivalent power NEP

Noise-equivalent power (NEP): incident optical power that produces a signal-to-noise of unity per square-root bandwidth at a given operating wavelength (λ_0) and modulation frequency (*f*)

$$i_{p,rms}(f) = \sigma_{r}(f,B)$$

$$\Re(\lambda_{0}) \cdot P_{eq,rms}(\lambda_{0},f) = \sqrt{\int_{f-B/2}^{f+B/2} S(f') df'}$$

$$\mathsf{NEP}(\lambda_0, f) = \frac{P_{\mathsf{eq,rms}}(\lambda_0, f)}{\sqrt{B}} = \frac{\sqrt{S}(f)}{\Re(\lambda_0)} \left(\frac{\mathsf{W}}{\sqrt{\mathsf{Hz}}}\right)$$

NEP of InSb photodiodes NEP $\propto \sqrt{A}$ NEP

Frequency (Hz)

resistance-limited receiver:

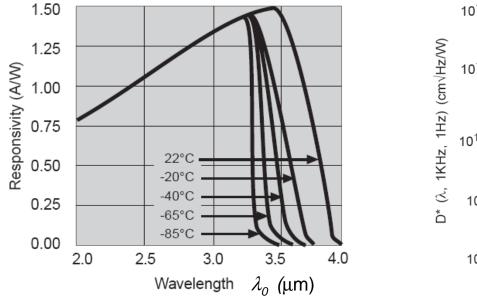
$$\overline{S}(f) = S = 2e \overline{Gi_{d}} F + \frac{4k_{B}T}{R_{L}}$$

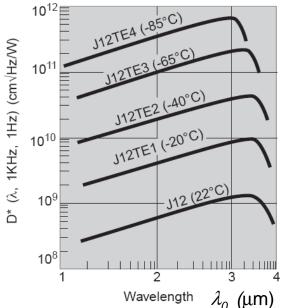
Detectivity D*

Detectivity (D^*) : reciprocal NEP per square-root area

$$D^* = \frac{\sqrt{A}}{NEP} = \Re(\lambda_0) \sqrt{\frac{A}{\overline{S}(f)}} \quad \left(\frac{\mathrm{cm}\sqrt{\mathrm{Hz}}}{\mathrm{W}} = \mathrm{Jones}\right)$$

Responsivity and detectivity of InAs photodiodes





Exercises

1. Consider three photodetectors in series with a 50- Ω load resistor that are to be used in detecting a 1- μ m wavelength optical signal with a bandwidth *B* = 1 GHz bandwidth at 77K (liquid nitrogen temperature):

- (i) a *p-i-n* photodiode with quantum efficiency $\eta = 0.9$;
- (ii) an avalanche photodiode (APD) with quantum efficiency $\eta = 0.6$, mean gain G = 100 and excess noise factor F = 2;
- (iii) a 10-stage photomultiplier tube (PMT) with quantum efficiency $\eta = 0.3$, overall mean gain $G = 4^{10}$ and overall variance $\sigma_{G}^2 = G^2/4$.
- a) For each detector, find the photocurrent signal-to-noise ratio when the detector is illuminated by a photon flux $\Phi = 10^{10} \text{ s}^{-1}$.
- b) By which devices is the signal detectable?
- 2. Give an estimate for *NEP* and *D*^{*} of a InGaAs photomultipier detecting 1.55 μ m radiation, with the following parameters:
 - Cathode quantum efficiency $\eta = 8 \times 10^{-3}$
 - Cathode area $A = 3 \times 8 \text{ mm}^2$
 - Gain $G = 1 \times 10^{6}$
 - Anode dark current $I_d = 40$ nA